

**STATISTICS OF SIGNAL TO INTERFERENCE RATIO PROCESS
AT OUTPUT OF MOBILE-TO-MOBILE RAYLEIGH FADING
CHANNEL IN THE PRESENCE
OF COCHANNEL INTERFERENCE**

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Abstract. *Dual-hop cooperative communications in interference-limited Rayleigh fading channel are investigated in this paper. The paper considers the first- and second-order statistics of the signal to interference ratio process at the input of the destination mobile station. The exact closed-form expressions for the first-order statistical measures, the probability density function and cumulative distribution function, are derived. We also derive the approximate closed form expressions for the second-order statistics, the level crossing rate and the average fade duration. The obtained theoretical results are verified by the Monte-Carlo simulations.*

Key words: *mobile-to-mobile channel, Rayleigh fading, probability density function, level crossing rate, average fade duration, cooperative communications*

1. INTRODUCTION

An efficient way to improve capacity, reliability and energy efficiency of mobile communications is to use the cooperation between users [1,2]. In cooperative communications, the mobile stations are connected either directly, or via a relay or both.

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The cooperative communications are especially important for mobile-to-mobile (M2M) channels, where both mobile stations are in motion and often there is no direct link between the source (S) and the destination (D). M2M channel will have even higher importance in the future, because of increasing importance of vehicle-to-vehicle (V2V) communications. In order to further increase the mobile system capacity, the frequency reuse is implemented. On the other hand, the frequency reuse causes strong cochannel interference (CCI). CCI is much stronger than the noise, and the noise may be neglected. Such an environment is called interference-limited.

Cooperative systems in Rayleigh fading channel are often analyzed in literature [3–13]. The outage probability, in the presence of Rayleigh interference, is considered in [3,4] when the desired signal is transmitted over Nakagami-m and Weibull fading channel, respectively. The average bit error rate, in the generalized K fading environment, is analyzed in [5]. Best relay selection cooperative communications, with the respect to the outage probability, are investigated in [6] for decode-and-forward (DF) and in [7,8] for amplify-and-forward (AF) relaying strategy. The outage performance of dual-hop cooperative systems in Rayleigh fading channels is studied for AF relays in [9–12] and for the case of DF relaying in [11]. Multihop relay systems with CCI in Rayleigh fading channel are evaluated in [13]. Second-order statistics, level crossing rate (LCR) and average fade duration (AFD), are analyzed in [14,15]. Paper [14] considers two-hop AF relaying in Rayleigh fading channel in the presence of thermal noise and CCI. On the other hand, paper [15] evaluates second-order statistical parameters for multiple hop Rayleigh fading channel, however in the absence of interference.

Having in mind the above analysis, it may be noticed that there is a lack of research of dual-hop cooperative systems in interference-limited Rayleigh fading channel, especially the investigation of the second-order statistics, LCR and AFD. Therefore, in this paper we derive the exact closed-form expressions for the first-order statistics of the signal to interference ratio at the input of the destination mobile station, probability density function (PDF) and cumulative distribution function (CDF) with AF relaying in mobile-to-mobile channel. The Laplace approximation for the LCR and AFD is also derived. The theoretical results are verified with the Monte-Carlo simulation. The analysis is also valid for mobile-to-base station channel, without the direct link between S and D. Besides, first order statistics with static transmitters and receivers may also be determined by the given analysis.

The paper is structured as follows. Section 2 describes the system model. First- and second-order performance measures are derived in Sections 3 and 4, respectively. Some numerical results, which show the influence of the fading channel parameters on the system's performance, are given in Section 5. The concluding remarks are given in Section 6.

2. SYSTEM MODEL

As already mentioned, we consider a dual-hop cooperative relay system, as shown in Fig. 1. Due to obstacles, there is no line-of-sight between the source and destination, and therefore S and D are connected only via a relay (R). The interference is present at both sections, S-R and R-D. Such a scenario is very likely for M2M communications, because

of the movement of both S and D. On the other hand, in cellular mobile-to-base station communications, there is usually also a direct link between S and D, due to high altitude location of the base station.

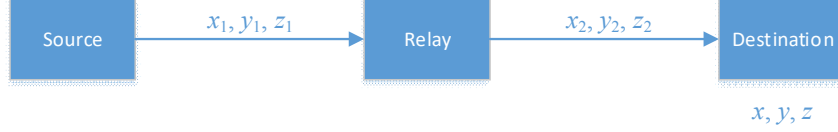


Fig. 1 System model

Random variables, describing the desired signal x_1, x_2 and interference y_1, y_2 envelopes, at the first and second section, are independent non-identically distributed Rayleigh random variables with the following probability density functions:

$$\begin{aligned}
 p_{x_i}(x_i) &= \frac{x_i}{\sigma_{x_i}^2} \cdot \exp\left(-\frac{x_i^2}{2\sigma_{x_i}^2}\right), \quad x_i \geq 0, \quad i = 1, 2, \\
 p_{y_i}(y_i) &= \frac{y_i}{\sigma_{y_i}^2} \cdot \exp\left(-\frac{y_i^2}{2\sigma_{y_i}^2}\right), \quad y_i \geq 0, \quad i = 1, 2,
 \end{aligned} \tag{1}$$

where $\sigma_{x_1}, \sigma_{x_2}, \sigma_{y_1}$ and σ_{y_2} are Rayleigh fading parameters. The signal to interference ratio is z_1, z_2 and z .

3. PDF, CDF AND OUTAGE PROBABILITY

The desired received signal envelope at the input of the destination mobile station is given by $x = x_1 \cdot x_2$. Similarly, the cochannel interference envelope at the input of the destination mobile station is equal to $y = y_1 \cdot y_2$. Since we consider AF relays, the signal to interference ratio at the input of the destination mobile station can be expressed as the ratio [16]

$$z = \frac{x}{y} = \frac{x_1 \cdot x_2}{y_1 \cdot y_2} = z_1 \cdot z_2, \tag{2}$$

where $z_1 = x_1 / y_1$ and $z_2 = x_2 / y_2$. We will first derive PDF of z_1 and z_2 , and finally of $z = z_1 \cdot z_2$. Since $z_i, i = 1, 2$ is the ratio of the random variables x_i and y_i , its PDF is defined as

$$\begin{aligned}
p_{z_i}(z_i) &= \int_0^{\infty} dy_i \cdot y_i \cdot p_{x_i}(y_i \cdot z_i) \cdot p_{y_i}(y_i) \\
&= \int_0^{\infty} dy_i \cdot y_i \cdot \frac{y_i z_i}{\sigma_{x_i}^2} \cdot \exp\left(-\frac{y_i^2 z_i^2}{2\sigma_{x_i}^2}\right) \cdot \frac{y_i}{\sigma_{y_i}^2} \cdot \exp\left(-\frac{y_i^2}{2\sigma_{y_i}^2}\right) \\
&= \frac{z_i}{\sigma_{x_i}^2 \sigma_{y_i}^2} \int_0^{\infty} dy_i \cdot y_i^3 \cdot \exp\left(-y_i^2 \left(\frac{z_i^2}{2\sigma_{x_i}^2} + \frac{1}{2\sigma_{y_i}^2}\right)\right) \\
&= \frac{z_i}{\sigma_{x_i}^2 \sigma_{y_i}^2} \int_0^{\infty} ds \cdot s \cdot \exp\left(-s \left(\frac{z_i^2}{2\sigma_{x_i}^2} + \frac{1}{2\sigma_{y_i}^2}\right)\right).
\end{aligned} \tag{3}$$

After integrating by parts, we get

$$p_{z_i}(z_i) = \frac{8\sigma_{y_i}^2 \sigma_{x_i}^2 \cdot z_i}{\left(2\sigma_{x_i}^2 + 2\sigma_{y_i}^2 \cdot z_i^2\right)^2}. \tag{4}$$

Now, we have

$$p_z(z) = \int_0^{\infty} dz_2 \frac{1}{z_2} p_{z_1}(z/z_2) p_{z_2}(z_2). \tag{5}$$

After substituting (3) in (5), we get

$$p_z(z) = \frac{4 \cdot \sigma_{x_1}^2 \cdot \sigma_{y_2}^2}{\sigma_{y_1}^2 \cdot \sigma_{x_2}^2} \cdot \frac{1}{z^3} \cdot \int_0^{\infty} dz_2 \cdot \frac{z_2^3}{\left(1 + \frac{\sigma_{x_1}^2}{\sigma_{y_1}^2 z^2} z_2^2\right)^2 \left(1 + \frac{\sigma_{y_2}^2}{\sigma_{x_2}^2} z_2^2\right)^2}. \tag{6}$$

Using 3.259.3 [17], the final expression for the destination SIR PDF may be obtained as

$$p_z(z) = \frac{1}{3} \cdot \frac{\sigma_{y_1}^2 \cdot \sigma_{y_2}^2}{\sigma_{x_1}^2 \cdot \sigma_{x_2}^2} \cdot z \cdot {}_2F_1\left(2, 2; 4; 1 - z^2 \frac{\sigma_{y_1}^2 \cdot \sigma_{y_2}^2}{\sigma_{x_1}^2 \cdot \sigma_{x_2}^2}\right), \tag{7}$$

where ${}_2F_1(a, b; c; d)$ is Hypergeometric function [17].

The cumulative distribution function of z is defined as

$$F_z(z) = \int_0^z dt \cdot p_z(t). \tag{8}$$

By inserting (6) in (8), and after some mathematical manipulations we get

$$F_z(z) = 2 \cdot \frac{\sigma_{y_2}^2}{\sigma_{x_2}^2} \cdot \int_0^\infty dz_2 \cdot \frac{z_2}{\left(1 + \frac{\sigma_{x_1}^2}{\sigma_{y_1}^2} z_2^2\right) \left(1 + \frac{\sigma_{y_2}^2}{\sigma_{x_2}^2} z_2^2\right)^2}. \quad (9)$$

Again, using 3.259.3 [17], we obtain the final expression

$$F_z(z) = \frac{1}{2} \cdot \frac{\sigma_{y_1}^2 \cdot \sigma_{y_2}^2}{\sigma_{x_1}^2 \cdot \sigma_{x_2}^2} \cdot z^2 \cdot {}_2F_1\left(2, 2; 3; 1 - z^2 \frac{\sigma_{y_1}^2 \cdot \sigma_{y_2}^2}{\sigma_{x_1}^2 \cdot \sigma_{x_2}^2}\right). \quad (10)$$

The outage probability, defined as the probability that the signal to interference ratio at the destination mobile station is lower than a certain threshold z_{th} is equal to

$$P_{out}(z_{th}) = \Pr[z < z_{th}] = F_z(z_{th}). \quad (11)$$

4. LEVEL CROSSING RATE AND AVERAGE FADE DURATION

The destination mobile station signal to interference ratio LCR is evaluated as the average value of the first derivative of the SIR.

Having in mind (2), the first derivative of z is

$$\dot{z} = \frac{x_2}{y_1 \cdot y_2} \dot{x}_1 + \frac{x_1}{y_1 \cdot y_2} \dot{x}_2 - \frac{x_1 \cdot x_2}{y_1^2 \cdot y_2} \dot{y}_1 - \frac{x_1 \cdot x_2}{y_1 \cdot y_2^2} \dot{y}_2. \quad (12)$$

Random variables \dot{x}_1 , \dot{x}_2 , \dot{y}_1 , and \dot{y}_2 have zero-mean Gaussian distribution with the following variances [18]:

$$\begin{aligned} \sigma_{\dot{x}_i}^2 &= 2\pi^2 f_m^2 \sigma_{x_i}^2, \\ \sigma_{\dot{y}_i}^2 &= 2\pi^2 f_m^2 \sigma_{y_i}^2, \quad i = 1, 2. \end{aligned} \quad (13)$$

where f_m is the maximum Doppler frequency.

Since a linear transformation of a Gaussian random variable is also a Gaussian random variable, \dot{z} follows a conditional Gaussian distribution with mean [15]

$$\bar{\dot{z}} = \frac{x_2}{y_1 \cdot y_2} \bar{\dot{x}_1} + \frac{x_1}{y_1 \cdot y_2} \bar{\dot{x}_2} - \frac{x_1 \cdot x_2}{y_1^2 \cdot y_2} \bar{\dot{y}_1} - \frac{x_1 \cdot x_2}{y_1 \cdot y_2^2} \bar{\dot{y}_2} = 0, \quad (14)$$

and variance

$$\sigma_{\dot{z}}^2 = 2\pi^2 f_m^2 \frac{x_2^2}{y_1^2 \cdot y_2^2} \sigma_{x_1}^2 \left(1 + z^2 \frac{y_1^2 \cdot y_2^2}{x_2^4} \frac{\sigma_{x_2}^2}{\sigma_{x_1}^2} + z^2 \frac{y_2^2}{x_2^2} \frac{\sigma_{y_1}^2}{\sigma_{x_1}^2} + z^2 \frac{y_1^2}{x_2^2} \frac{\sigma_{y_2}^2}{\sigma_{x_1}^2}\right). \quad (15)$$

The joint probability density function of z, \dot{z}, x_2, y_1 , and y_2 is

$$p_{z\dot{z}x_2y_1y_2}(z, \dot{z}, x_2, y_1, y_2) = p_{\dot{z}}(\dot{z} | z, x_2, y_1, y_2) \cdot p_{zx_2y_1y_2}(z, x_2, y_1, y_2). \quad (16)$$

where

$$p_{zx_2y_1y_2}(z, x_2, y_1, y_2) = p_{x_2}(x_2) \cdot p_{y_1}(y_1) \cdot p_{y_2}(y_2) \cdot p_z(z | x_2, y_1, y_2). \quad (17)$$

The conditional joint probability density function of z is

$$p_z(z | x_2, y_1, y_2) = \left| \frac{dx_1}{dz} \right| p_{x_1} \left(\frac{z \cdot y_1 \cdot y_2}{x_2} \right). \quad (18)$$

From (2) we get

$$\left| \frac{dx_1}{dz} \right| = \frac{y_1 \cdot y_2}{x_2}. \quad (19)$$

After substituting (17), (18), and (19) into (16), the expression for $p_{zzx_2y_1y_2}(z, \dot{z}, x_2, y_1, y_2)$ becomes

$$p_{zzx_2y_1y_2}(z, \dot{z}, x_2, y_1, y_2) = \frac{y_1 \cdot y_2}{x_2} p_{x_1} \left(\frac{z \cdot y_1 \cdot y_2}{x_2} \right) \cdot p_{x_2}(x_2) \cdot p_{y_1}(y_1) \cdot p_{y_2}(y_2) \cdot p_z(\dot{z} | z, x_2, y_1, y_2). \quad (20)$$

The joint probability density function of z and \dot{z} is

$$p_{zz}(z, \dot{z}) = \int_0^\infty dx_2 \int_0^\infty dy_1 \int_0^\infty dy_2 \cdot p_{zzx_2y_1y_2}(z, \dot{z}, x_2, y_1, y_2). \quad (21)$$

From (20) and (21), we get

$$p_{zz}(z, \dot{z}) = \int_0^\infty dx_2 \int_0^\infty dy_1 \int_0^\infty dy_2 \cdot \frac{y_1 \cdot y_2}{x_2} p_{x_1} \left(\frac{z \cdot y_1 \cdot y_2}{x_2} \right) \cdot p_{x_2}(x_2) \cdot p_{y_1}(y_1) \cdot p_{y_2}(y_2) \cdot p_z(\dot{z} | z, x_2, y_1, y_2). \quad (22)$$

Finally, the level crossing rate of signal to interference ratio process at the input of the destination mobile station is [19]

$$\begin{aligned} N_z(z) &= \int_0^\infty d\dot{z} \cdot \dot{z} \cdot p_{zz}(z\dot{z}) \\ &= \int_0^\infty dx_2 \int_0^\infty dy_1 \int_0^\infty dy_2 \cdot \frac{y_1 \cdot y_2}{x_2} p_{x_1} \left(\frac{z \cdot y_1 \cdot y_2}{x_2} \right) \cdot p_{x_2}(x_2) \cdot p_{y_1}(y_1) \cdot p_{y_2}(y_2) \\ &\quad \cdot \int_0^\infty d\dot{z} \cdot \dot{z} \cdot p_z(\dot{z} | z, x_2, y_1, y_2). \end{aligned} \quad (23)$$

Having in mind that

$$\int_0^\infty d\dot{z} \cdot \dot{z} \cdot p_z(\dot{z} | z, x_2, y_1, y_2) = \frac{1}{\sqrt{2\pi}\sigma_z} \int_0^\infty \dot{z} \cdot e^{-\frac{\dot{z}^2}{2\sigma_z^2}} d\dot{z} = \frac{\sigma_z}{\sqrt{2\pi}}, \quad (24)$$

we have

$$N_z(z) = \int_0^\infty dx_2 \int_0^\infty dy_1 \int_0^\infty dy_2 \cdot \frac{y_1 \cdot y_2}{x_2} p_{x_1} \left(\frac{z \cdot y_1 \cdot y_2}{x_2} \right) \cdot p_{x_2}(x_2) \cdot p_{y_1}(y_1) \cdot p_{y_2}(y_2) \cdot \frac{\sigma_z}{\sqrt{2\pi}}. \quad (25)$$

After substituting the probability density functions (1) in (25), the expression for the level crossing rate becomes

$$\begin{aligned} N_z(z) &= \frac{1}{\sigma_{x_1} \cdot \sigma_{x_2}^2 \cdot \sigma_{y_1}^2 \cdot \sigma_{y_2}^2} \cdot z \\ &\cdot \sqrt{\pi} \cdot f_m \cdot \int_0^\infty dx_2 \int_0^\infty dy_1 \int_0^\infty dy_2 \cdot \\ &\cdot \sqrt{1 + z^2 \frac{y_1^2 \cdot y_2^2 \cdot \sigma_{x_2}^2}{x_2^4 \cdot \sigma_{x_1}^2} + z^2 \frac{y_2^2 \cdot \sigma_{y_1}^2}{x_2^2 \cdot \sigma_{x_1}^2} + z^2 \frac{y_1^2 \cdot \sigma_{y_2}^2}{x_2^2 \cdot \sigma_{x_1}^2}} \\ &\cdot \exp \left(-\frac{1}{2\sigma_{x_1}^2} \frac{z^2 \cdot y_1^2 \cdot y_2^2}{x_2^2} - \frac{1}{2\sigma_{x_2}^2} x_2^2 - \frac{1}{2\sigma_{y_1}^2} y_1^2 - \frac{1}{2\sigma_{y_2}^2} y_2^2 + 2 \ln y_1 + 2 \ln y_2 \right). \end{aligned} \quad (26)$$

The three-fold integral in the above expression is solved by using Laplace approximation theorem for the three-fold integral [20]

$$\begin{aligned} &\int_0^\infty dx \int_0^\infty dy \int_0^\infty dz \cdot g(x, y, z) \exp(-\lambda \cdot f(x, y, z)) \\ &= \left(\frac{2\pi}{\lambda} \right)^{3/2} \frac{g(x_0, y_0, z_0)}{\sqrt{B(x_0, y_0, z_0)}} \exp(-\lambda \cdot f(x_0, y_0, z_0)), \end{aligned} \quad (27)$$

where $x_0, y_0,$ and z_0 are solutions of the following set of equations:

$$\begin{aligned} \frac{\partial f(x_0, y_0, z_0)}{\partial x_0} &= 0, \\ \frac{\partial f(x_0, y_0, z_0)}{\partial y_0} &= 0, \\ \frac{\partial f(x_0, y_0, z_0)}{\partial z_0} &= 0, \end{aligned} \quad (28)$$

and

$$B(x_0, y_0, z_0) = \begin{vmatrix} \frac{\partial^2 f(x_0, y_0, z_0)}{\partial x_0^2} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial x_0 \partial y_0} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial y_0 \partial z_0} \\ \frac{\partial^2 f(x_0, y_0, z_0)}{\partial x_0 \partial y_0} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial y_0^2} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial y_0 \partial z_0} \\ \frac{\partial^2 f(x_0, y_0, z_0)}{\partial x_0 \partial z_0} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial y_0 \partial z_0} & \frac{\partial^2 f(x_0, y_0, z_0)}{\partial z_0^2} \end{vmatrix}. \quad (29)$$

For the considered case, the constant $\lambda = 1$ and functions f and g are

$$g(x_2, y_1, y_2) = \sqrt{1 + z^2 \frac{y_1^2 \cdot y_2^2 \sigma_{x_2}^2}{x_2^4 \sigma_{x_1}^2} + z^2 \frac{y_2^2 \sigma_{y_1}^2}{x_2^2 \sigma_{x_1}^2} + z^2 \frac{y_1^2 \sigma_{y_2}^2}{x_2^2 \sigma_{x_1}^2}}, \quad (30)$$

$$f(x_2, y_1, y_2) = \frac{1}{2\sigma_{x_1}^2} \frac{z^2 \cdot y_1^2 \cdot y_2^2}{x_2^2} + \frac{1}{2\sigma_{x_2}^2} x_2^2 + \frac{1}{2\sigma_{y_1}^2} y_1^2 + \frac{1}{2\sigma_{y_2}^2} y_2^2 - 2 \ln y_1 - 2 \ln y_2. \quad (31)$$

Using the expressions for CDF and LCR, the average fade duration may be defined as

$$T_z(z) = \frac{F_z(z)}{N_z(z)}. \quad (32)$$

5. NUMERICAL RESULTS

This section presents some numerical results that indicate the influence of different fading parameters on the PDF, outage probability, LCR, and AFD. The obtained theoretical results are confirmed by the Monte-Carlo simulation, with one million simulation steps. Without the loss of generality, the following assumptions are made $\sigma_{x_1} = \sigma_{x_2} = \sigma_x$ and $\sigma_{y_1} = \sigma_{y_2} = \sigma_y$.

Fig. 2 shows PDF of the destination node signal to interference ratio, for different values of ratio $R = 20 \log(\sigma_x / \sigma_y)$. The results show an excellent agreement between the theoretical and simulation results.

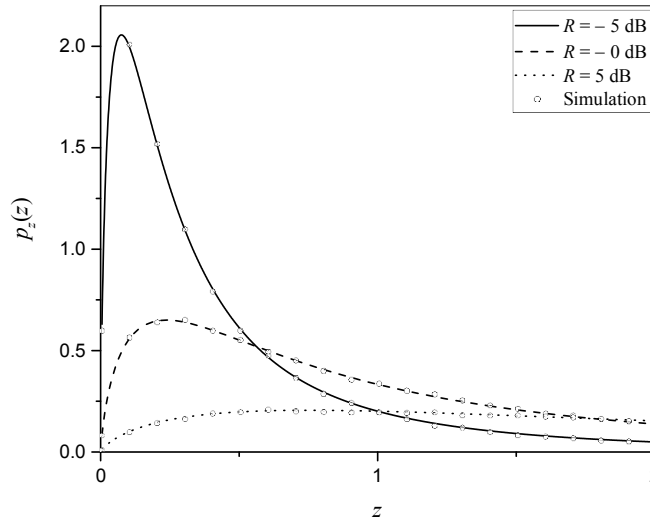


Fig. 2 Probability density function of the destination signal to interference ratio

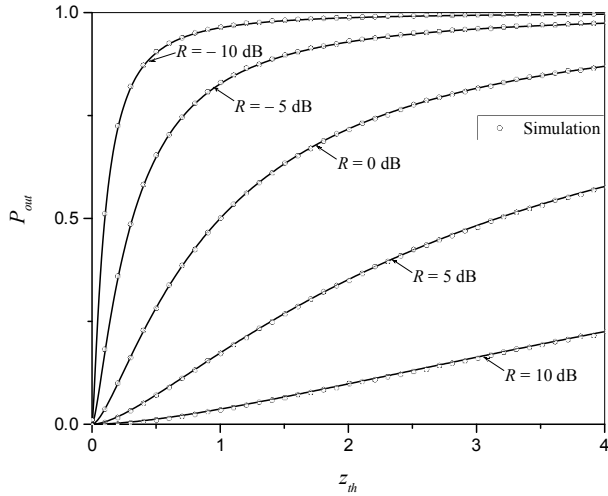


Fig. 3 Outage probability as a function of the outage threshold

Outage probability, P_{out} , as a function of the outage threshold z_{th} is shown in Fig. 3, with R as a parameter. It may be noticed that P_{out} is lower for higher R , due to better channel conditions for higher signal to interference ratio. Again, the theoretical and simulation results are in good agreement.

Fig. 4 illustrates the outage probability as a function of R . This figure confirms conclusions from the Fig. 3.

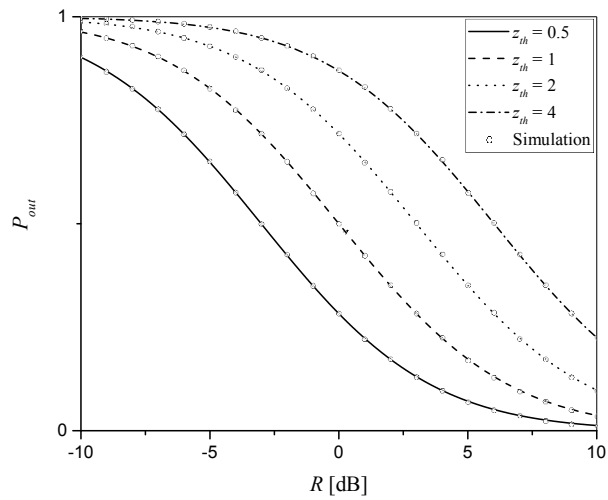


Fig. 4 Outage probability as a function of average signal to interference power ratio

Fig. 5 depicts the normalized level crossing rate for different values of R . Besides the Laplace approximation results, we show the results obtained by the numerical integration of (26) in the software package Mathematica. Also, these results are compared to the Monte-Carlo simulation results, based on the sum-of-sinusoids Rayleigh channel model [21]. The carrier frequency is chosen to be 1 GHz, mutual terminals speed is 80 km/h, which resulted in the maximum Doppler frequency of 74 Hz. There is a good agreement between the exact (numerical integration), approximation, and simulation results. The same difference between the theoretical and simulation results for LCR of a Rayleigh random variable is observed in [21] (Fig. 9), too.

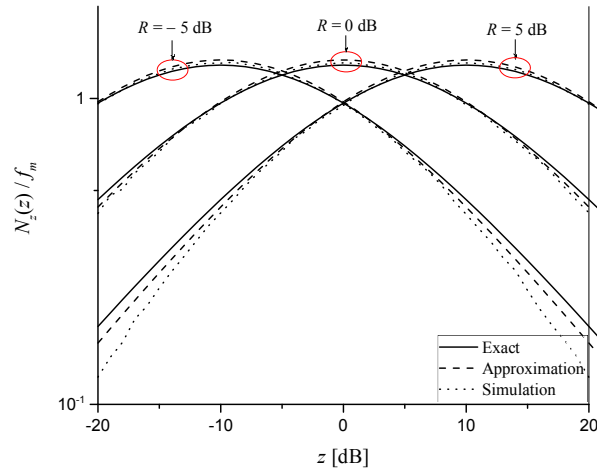


Fig. 5 Normalized level crossing rate

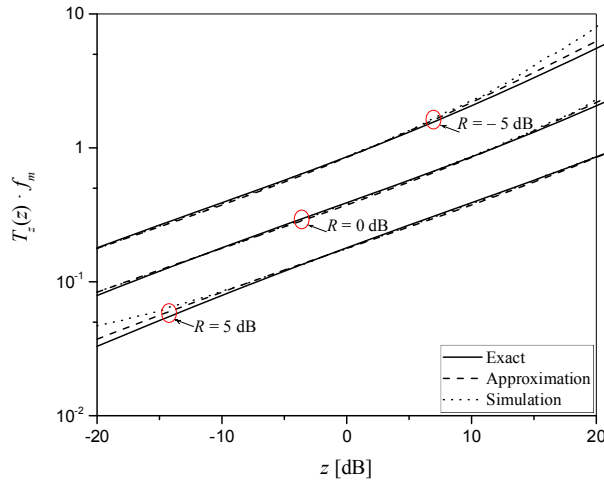


Fig. 6 Normalized average fade duration

Based on the level crossing rate and the cumulative distribution function, the normalized average fade duration is calculated and shown in Fig. 6. The exact, Laplace approximation and the simulation results are shown. The average fade duration is lower for higher R , again due to better channel conditions.

6. CONCLUSION

This paper considers the cooperative mobile-to-mobile communications performance in Rayleigh fading channel in the presence of cochannel interference. Two hop communication is assumed, where source mobile station is connected to the destination mobile station via a relay. Both the desired signal and interference are subject to Rayleigh fading. The exact closed-form of the first-order statistical measures, the probability density function and cumulative distribution function are derived. Besides, an approximate closed-form expression for the level crossing rate and average fade duration is given. The exact and approximate results are compared to the Monte-Carlo simulation results. The analysis shows an excellent agreement between the exact, approximate and simulated results.

In the future work, we will analyze more complex system model, where the source and destination are connected both directly and via a relay.

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REFERENCES

- [1] J. N. Laneman, D. N. C. Tse, and G. W. Wornell, "Cooperative diversity in wireless networks: Efficient protocols and outage behavior," *IEEE Trans. Inf. Theory*, vol. 50, no. 12, pp. 3062–3080, 2004. DOI:10.1109/TIT.2004.838089.
- [2] B. Talha and M. Pätzold, "Channel Models for Mobile-to-Mobile Cooperative Communication Systems: A State of the Art Review," *IEEE Veh. Technol. Mag.*, vol. 6, no. 2, pp. 33–43, 2011. DOI:10.1109/MVT.2011.940793.
- [3] K. P. Peppas, G. P. Eftymoglou, and V. A. Aalo, "Exact and asymptotic analysis of dual-hop AF systems in Nakagami-m fading with Rayleigh interferers," in *Proceeding of the 2016 European Conference on Networks and Communications (EuCNC)*, Athens, Greece, 2016, pp. 288–292.
- [4] D. N. Milić, D. B. Djošić, Č. M. Stefanović, M. M. Smilić, and S. N. Suljović, "Outage performance of multi-branch SC receiver over correlated Weibull channel in the presence of correlated Rayleigh cochannel interference," *Facta Univ. Ser. Autom. Control Robot.*, vol. 14, no. 3, pp. 183–191, 2015.
- [5] J. Anastasov and A. Cvetković, "Error probability evaluation of multiuser wireless system over generalized fading environment," *Facta Univ. Ser. Autom. Control Robot.*, vol. 1, no. 2, pp. 137–146, 2016.
- [6] T. T. Duy, G. C. Alexandropoulos, V. T. Tung, V. N. Son, and T. Q. Duong, "Outage performance of cognitive cooperative networks with relay selection over double-Rayleigh fading channels," *IET Commun.*, vol. 10, no. 1, pp. 57–64, 2016. DOI:10.1049/iet-com.2015.0236.
- [7] V. N. Q. Bao, T. Q. Duong, D. B. da Costa, G. C. Alexandropoulos, and A. Nallanathan, "Cognitive amplify-and-forward relaying with best relay selection in non-identical Rayleigh fading," *IEEE Commun.*

- Lett., vol. 17, no. 3, pp. 475–478, 2013. DOI:10.1109/LCOMM.2013.011513.122213.
- [8] S. S. Ikki and M. H. Ahmed, “Performance of multiple-relay cooperative diversity systems with best relay selection over Rayleigh fading channels,” *EURASIP J. Adv. Signal Process.*, vol. 2008, p. 145, 2008. DOI:10.1155/2008/580368.
- [9] W. Xu, J. Zhang, and P. Zhang, “Outage Probability of Two-Hop Fixed-Gain Relay with Interference at the Relay and Destination,” *IEEE Commun. Lett.*, vol. 15, no. 6, pp. 608–610, 2011. DOI:10.1109/LCOMM.2011.040711.102016.
- [10] A. M. Cvetkovic, G. T. Djordjevic, and M. Č. Stefanović, “Performance of interference-limited dual-hop non-regenerative relays over Rayleigh fading channels,” *IET Commun.*, vol. 5, no. 2, p. 135–140(5), 2011. DOI: 10.1049/iet-com.2010.0019.
- [11] D. Lee and J. H. Lee, “Outage probability for dual-hop relaying systems with multiple interferers over Rayleigh fading channels,” *IEEE Trans. Veh. Technol.*, vol. 60, no. 1, pp. 333–338, 2011. DOI:10.1109/TVT.2010.2089998.
- [12] H. A. Suraweera, H. K. Garg, and A. Nallanathan, “Performance Analysis of Two Hop Amplify-and-Forward Systems with Interference at the Relay,” *IEEE Commun. Lett.*, vol. 14, no. 8, pp. 692–694, 2010. DOI:10.1109/LCOMM.2010.08.100109.
- [13] T. Soithong, V. A. Aalo, G. P. Eftymoglou, and C. Chayawan, “Performance of Multihop Relay Systems with Co-Channel Interference in Rayleigh Fading Channels,” *IEEE Commun. Lett.*, vol. 15, no. 8, pp. 836–838, 2011. DOI:10.1109/LCOMM.2011.062211.110747.
- [14] M. S. Gilan, M. Y. Manesh, and A. Mohammadi, “Level Crossing Rate and Average Fade Duration of Amplify and Forward Relay Channels with Cochannel Interference,” in *Proceedings of the 22th European Wireless Conference*, Oulu, Finland, 2016, pp. 1–5.
- [15] Z. Hadzi-Velkov, N. Zlatanov, and G. K. Karagiannidis, “On the second order statistics of the multihop rayleigh fading channel,” *IEEE Trans. Commun.*, vol. 57, no. 6, pp. 1815–1823, 2009. DOI:10.1109/TCOMM.2009.06.070460.
- [16] M. D. Yacoub, J. V. Bautista, and L. G. de Rezende Guedes, “On higher order statistics of the Nakagami-m distribution,” *IEEE Trans. Veh. Technol.*, vol. 48, no. 3, pp.790-794, 1999. DOI: 10.1109/25.764995.
- [17] E. S. Gradstein and I. M. Ryzhik, *Table of integrals, sums, series, and products*, Seventh ed. San Diego, CA, USA: Academic Press, 2007.
- [18] M. D. Yacoub, J. E. V. Bautista, and L. Guerra de Rezende Guedes, “On higher order statistics of the Nakagami-m distribution,” *IEEE Trans. Veh. Technol.*, vol. 48, no. 3, pp. 790–794, 1999. DOI:10.1109/25.764995.
- [19] W. Jakes, *Microwave Mobile Communications*, 2nd ed. Piscataway, NJ: IEEE Press, 1993.
- [20] R. Wong, *Asymptotic approximations of integrals*. SIAM, 2001.
- [21] Y. R. Zheng and C. Xiao, “Simulation models with correct statistical properties for Rayleigh fading channels,” *IEEE Trans. Commun.*, vol. 51, no. 6, pp. 920–928, 2003. DOI:10.1109/TCOMM.2003.813259